

■GENERAL DESCRIPTION

The MYRGP-W/MYRGP-B series is a synchronous step-down mini DC/DC converter Which integrates an inductor and a control IC in one tiny package (2.0mm×2.5mm, h=1.0mm). An internal coil simplifies the circuit and enables minimization of noise and other operational Trouble due to the circuit wiring. A wide operating voltage range of 2.5V to 5.5V enables Support for applications that require an internally fixed output voltage (0.8V to 3.6V). The MYRGP-W/MYRGP-B series use synchronous rectification at an operating frequency of 3.0MHz. The MYRGP-W/MYRGP-B series use High Speed Transient Response -COT synchronous rectification.

PWM control (MYRGP-W) or PWM/PFM switching control (MYRGP-B) automatic can be selected.

The series have a high speed soft-start as fast as 0.3ms in typical for quick turn-on. With the built-in UVLO (Under Voltage LockOut) function, the internal P-channel driver transistor is forced OFF when input voltage becomes 2.0V or lower. When CE=Low, the integrated C_L discharge function which enables the electric charge at the output capacitor C_L to be discharged via the internal discharge switch located between the L_x and V_{SS} pins. The power consumption will be less than 1.0μA.



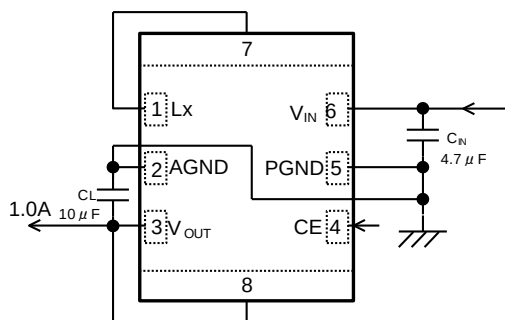
■APPLICATIONS

- Note Book PC
- Tablet PC
- SSD(Solid State Drive)
- Mobile phone
- Digital still camera
- Portable game machine

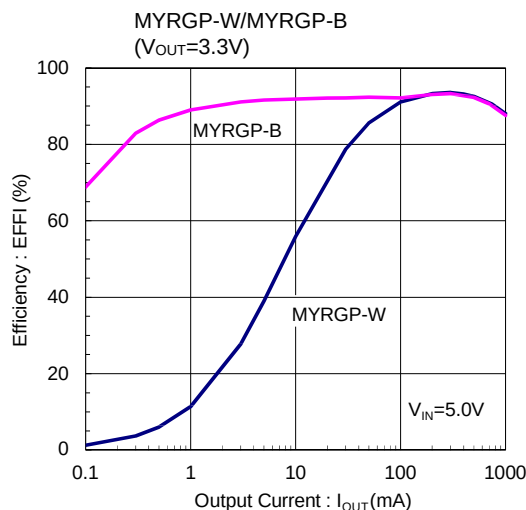
■FEATURES

Input Voltage	: 2.5V~5.5V
Output Voltage	: 0.8V~3.6V(±2.0%)
Switching Frequency	: 3.0MHz
Output Current	: 1.0A
Efficiency	: 93% (V _{IN} =5.0V, V _{OUT} =3.3V/300mA)
Control Methods	: High Speed Transient Response -COT 100% Duty Cycle PWM (MYRGP-W) PWM/PFM (MYRGP-B)
Circuit Protection	: Thermal Shut Down Current Limit Circuit (Drop) Short Circuit Protection (Latch)
Functions	: Soft-start Circuit Built-in UVLO C _L Discharge
Output Capacitor	: Low ESR Ceramic Capacitor
Operating Ambient Temperature	: -40°C~+105°C
Environmentally Friendly	: EU RoHS Compliant, Pb Free Halogen Free

■TYPICAL APPLICATION CIRCUIT

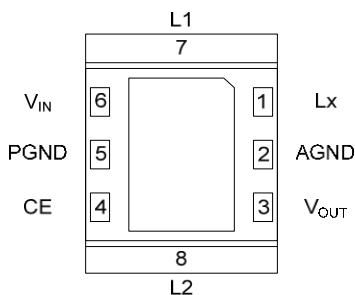


■TYPICAL PERFORMANCE CHARACTERISTICS



Diodes inside the circuit are an ESD protection diode and a parasitic diode.

PIN CONFIGURATION



BOTTOM VIEW

- * It should be connected the pin No.2 and 5 to the GND pin.
- * If the dissipation pad needs to be connected to other pins, it should be connected to the GND pin.
- * Please refer to pattern layout page for the connecting to PCB.

PIN ASSIGNMENT

PIN NUMBER	PIN NAME	FUNCTIONS
1	Lx	Switching Output
2	AGND	Analog Ground
3	V _{OUT}	Fixed Output Voltage PIN
4	CE	Chip Enable
5	PGND	Power Ground
6	V _{IN}	Power Input
7	L1	Inductor Electrodes
8	L2	Inductor Electrodes

FUNCTION TABLE

CE PIN Function

PIN NAME	SIGNAL	STATUS
CE	Low	Stand-by
	High	Active

* Please do not leave the CE pin open.

ABSOLUTE MAXIMUM RATINGS

Ta=25°C

PARAMETER	SYMBOL	RATINGS	UNITS
V _{IN} Pin Voltage	V _{IN}	-0.3 ~ +6.2	V
L _x Pin Voltage	V _{Lx}	- 0.3 ~ V _{IN} + 0.3 or + 6.2 ^{(*)1}	V
V _{OUT} Pin Voltage	V _{out}	- 0.3 ~ V _{IN} + 0.3 or + 4.0 ^{(*)2}	V
CE Pin Voltage	V _{CE}	- 0.3 ~ +6.2	V
Power Dissipation	P _d	1000 ^{(*)3}	mW
Operating Ambient Temperature	T _{opr}	- 40 ~ +105	°C
Storage Temperature	T _{stg}	- 55 ~ +125	°C

All voltages are described based on the GND (AGND and PGND) pin.

^{(*)1} The maximum value should be either V_{IN}+0.3 or +6.2 in the lowest voltage

^{(*)2} The maximum value should be either V_{IN}+0.3 or +4.0 in the lowest voltage

^{(*)3} The power dissipation figure shown is PCB mounted (40mm×40mm, t=1.6mm, Glass Epoxy FR-4).

Please refer to page 11 for details.

■ ELECTRICAL CHARACTERISTICS

● MYRGP-W/MYRGP-B

Ta=25°C

PARAMETER	SYMBOL	CONDITIONS	MIN.	TYP.	MAX.	UNITS	CIRCUIT
Output Voltage	V_{OUT}	When connected to external componets, $I_{OUT}=30mA$	<E-1>	<E-2>	<E-3>	V	①
Operating Voltage Range	V_{IN}		2.5	-	5.5	V	①
Maximum Output Current	I_{OUTMAX}	When connected to external components, $V_{IN}=<C-1>$	1000	-	-	mA	①
UVLO Voltage ⁽²⁾	V_{UVLO}	$V_{OUT}=0.6V$, Voltage which Lx pin holding "L" level ⁽⁶⁾	1.35	2	2.48	V	③
Quiescent Current (MYRGP-B)	I_q	$V_{OUT}=V_{OUT(E)} \times 1.1V$	-	25	40	μA	②
Quiescent Current (MYRGP-W)	I_q	$V_{OUT}=V_{OUT(E)} \times 1.1V$	-	400	825	μA	②
Stand-by Current	I_{STB}	$V_{CE}=0V$	-	0	1	μA	②
Minimum ON time ⁽²⁾	t_{ONmin}	When connected to external components, $V_{IN}=V_{CE}=<C-1>$, $I_{OUT}=1mA$	<E-5>	<E-6>	<E-7>	ns	①
Thermal Shutdown	T_{TSD}		-	150	-	°C	①
Thermal shutdown Hysteresis	T_{HYS}		-	30	-	°C	①
LxSW "H" ON Resistance	R_{LXH}	$V_{OUT}=0.6V$, $I_{LX}=100mA$ ⁽³⁾	-	0.24	0.37	Ω	④
LxSW "L" ON Resistance ⁽⁴⁾	R_{LXL}	$V_{OUT}=V_{OUT(T)} \times 1.1V$, $I_{LX}=100mA$ ⁽³⁾	-	0.16	0.3	Ω	④
LxSW "H" Leakage Current	I_{LeakH}	$V_{IN}=5.5V$, $V_{CE}=0V$, $V_{OUT}=0V$, $V_{LX}=5.5V$	-	0	30	μA	⑤
LxSW "L" Leakage Current	I_{LeakL}	$V_{IN}=5.5V$, $V_{CE}=0V$, $V_{OUT}=0V$, $V_{LX}=0V$	-	0	1	μA	⑤
Current Limit ⁽⁵⁾	I_{LIMH}	$V_{OUT}=0.6V$, I_{LX} until Lx pin oscillates	1.3	1.5	2.5	A	⑥
Output Voltage Temperature Characteristics	$\Delta V_{OUT}/(V_{OUT} \cdot \Delta T_{opr})$	$I_{OUT}=30mA$ $-40^\circ C \leq T_{opr} \leq 85^\circ C$	-	±100	-	ppm/°C	①
CE"H" Voltage	V_{CEH}	$V_{OUT}=0.6V$, Applied voltage to V_{CE} , Voltage changes Lx to "H" level ⁽⁶⁾	1.4	-	5.5	V	③
CE"L" Voltage	V_{CEL}	$V_{FB}=0.6V$, Applied voltage to V_{CE} , Voltage changes Lx to "L" level ⁽⁶⁾	AGND	-	0.3	V	③
CE"H" Current	I_{CEH}	$V_{IN}=5.5V$, $V_{CE}=5.5V$, $V_{OUT}=0V$	-0.1	-	0.1	μA	⑤
CE"L" Current	I_{CEL}	$V_{IN}=5.5V$, $V_{CE}=0V$, $V_{OUT}=0V$	-0.1	-	0.1	μA	⑤
Soft-start Time	t_{SS}	After "H" is fed to CE, the time by when clocks are generated at Lx pin.	0.1	0.3	0.5	ms	③
Short Protection Threshold Voltage	V_{SHORT}	Sweeping V_{OUT} , V_{OUT} voltage which Lx becomes "L" level ⁽⁶⁾	0.17	0.27	0.37	V	③
C _L Discharge	R_{DCHG}	$V_{CE}=0V$, $V_{OUT}=4.0V$	50	210	300	Ω	⑦
Inductance	L	Test Freq.=1MHz	-	1	-	μH	-
Inductor Rated Current	I_{DC}	$\Delta T=+40deg$	-	1.3	-	A	-

Unless otherwise stated, $V_{IN}=5.0V$, $V_{CE}=5.0V$, $V_{OUT(T)}$ =Nominal Voltage

NOTE:

⁽¹⁾ When the difference between the input and the output is small, 100% duty might come up and internal control circuits keep P-ch driver turning on even though the output current is not so large.

If current is further pulled from this state, output voltage will decrease because of P-ch driver ON resistance.

⁽²⁾ Including UVLO detect voltage, hysteresis operating voltage range for UVLO release voltage.

⁽³⁾ $R_{LXH}=(V_{IN} - Lx \text{ pin measurement voltage}) / 100mA$, $R_{LXL}=Lx \text{ pin measurement voltage} / 100mA$

⁽⁴⁾ Design value for the MYRGP-B series

⁽⁵⁾ Current limit denotes the level of detection at peak of coil current.

⁽⁶⁾ "H"= $V_{IN} \sim V_{IN} - 1.2V$, "L"= $-0.1V \sim +0.1V$

■ ELECTRICAL CHARACTERISTICS

● SPEC Table (V_{OUT} , t_{ONmin})
MYRGP-W; PWM Control

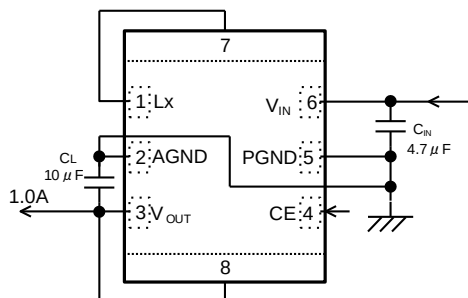
Part Number	NOMINAL OUTPUT VOLTAGE	V _{OUT}			t _{ONmin}				Part Number	NOMINA L OUTPUT VOLTAGE	V _{OUT}			t _{ONmin}			
					f _{OSC} =3.0MHz									f _{OSC} =3.0MHz			
		<E-1>	<E-2>	<E-3>	<C-1>	<E-5>	<E-6>	<E-7>			<C-1>	<E-5>	<E-6>	<E-7>			
	V _{OUT(T)}	MIN.	TYP.	MAX.	V _{IN}	MIN.	TYP.	MAX.		V _{OUT(T)}	MIN.	TYP.	MAX.	V _{IN}	MIN.	TYP.	MAX.
MYRGP 080100W21RA	0.80	0.784	0.800	0.816	2.70	71	119	166	MYRGP 225100W21RA	2.25	2.205	2.250	2.295	3.75	140	200	260
MYRGP 085100W21RA	0.85	0.833	0.850	0.867	2.70	72	121	169	MYRGP 230100W21RA	2.30	2.254	2.300	2.346	3.83	140	200	260
MYRGP 090100W21RA	0.90	0.882	0.900	0.918	2.70	73	122	171	MYRGP 235100W21RA	2.35	2.303	2.350	2.397	3.92	140	200	260
MYRGP 095100W21RA	0.95	0.931	0.950	0.969	2.70	74	123	172	MYRGP 240100W21RA	2.40	2.352	2.400	2.448	4.00	140	200	260
MYRGP 100100W21RA	1.00	0.980	1.000	1.020	2.70	86	123	160	MYRGP 245100W21RA	2.45	2.401	2.450	2.499	4.08	140	200	260
MYRGP 105100W21RA	1.05	1.029	1.050	1.071	2.70	91	130	169	MYRGP 250100W21RA	2.50	2.450	2.500	2.550	4.17	140	200	260
MYRGP 110100W21RA	1.10	1.078	1.100	1.122	2.70	95	136	177	MYRGP 255100W21RA	2.55	2.499	2.550	2.601	4.25	140	200	260
MYRGP 115100W21RA	1.15	1.127	1.150	1.173	2.70	99	142	185	MYRGP 260100W21RA	2.60	2.548	2.600	2.652	4.33	140	200	260
MYRGP 120100W21RA	1.20	1.176	1.200	1.224	2.70	104	148	193	MYRGP 265100W21RA	2.65	2.597	2.650	2.703	4.42	140	200	260
MYRGP 125100W21RA	1.25	1.225	1.250	1.275	2.70	108	154	201	MYRGP 270100W21RA	2.70	2.646	2.700	2.754	4.50	140	200	260
MYRGP 130100W21RA	1.30	1.274	1.300	1.326	2.70	112	160	209	MYRGP 275100W21RA	2.75	2.695	2.750	2.805	4.58	140	200	260
MYRGP 135100W21RA	1.35	1.323	1.350	1.377	2.70	117	167	217	MYRGP 280100W21RA	2.80	2.744	2.800	2.856	4.67	140	200	260
MYRGP 140100W21RA	1.40	1.372	1.400	1.428	2.70	121	173	225	MYRGP 285100W21RA	2.85	2.793	2.850	2.907	4.75	140	200	260
MYRGP 145100W21RA	1.45	1.421	1.450	1.479	2.70	125	179	233	MYRGP 290100W21RA	2.90	2.842	2.900	2.958	4.83	140	200	260
MYRGP 150100W21RA	1.50	1.470	1.500	1.530	2.70	130	185	241	MYRGP 295100W21RA	2.95	2.891	2.950	3.009	4.92	140	200	260
MYRGP 155100W21RA	1.55	1.519	1.550	1.581	2.70	134	191	249	MYRGP 300100W21RA	3.00	2.940	3.000	3.060	5.00	140	200	260
MYRGP 160100W21RA	1.60	1.568	1.600	1.632	2.70	138	198	257	MYRGP 305100W21RA	3.05	2.989	3.050	3.111	5.08	140	200	260
MYRGP 165100W21RA	1.65	1.617	1.650	1.683	2.75	140	200	260	MYRGP 310100W21RA	3.10	3.038	3.100	3.162	5.17	140	200	260
MYRGP 170100W21RA	1.70	1.666	1.700	1.734	2.83	140	200	260	MYRGP 315100W21RA	3.15	3.087	3.150	3.213	5.25	140	200	260
MYRGP 175100W21RA	1.75	1.715	1.750	1.785	2.92	140	200	260	MYRGP 320100W21RA	3.20	3.136	3.200	3.264	5.33	140	200	260
MYRGP 180100W21RA	1.80	1.764	1.800	1.836	3.00	140	200	260	MYRGP 325100W21RA	3.25	3.185	3.250	3.315	5.42	140	200	260
MYRGP 185100W21RA	1.85	1.813	1.850	1.887	3.08	140	200	260	MYRGP 330100W21RA	3.30	3.234	3.300	3.366	5.50	140	200	260
MYRGP 190100W21RA	1.90	1.862	1.900	1.938	3.17	140	200	260	MYRGP 335100W21RA	3.35	3.283	3.350	3.417	5.50	142	203	264
MYRGP 195100W21RA	1.95	1.911	1.950	1.989	3.25	140	200	260	MYRGP 340100W21RA	3.40	3.332	3.400	3.468	5.50	144	206	268
MYRGP 200100W21RA	2.00	1.960	2.000	2.040	3.33	140	200	260	MYRGP 345100W21RA	3.45	3.381	3.450	3.519	5.50	146	209	272
MYRGP 205100W21RA	2.05	2.009	2.050	2.091	3.42	140	200	260	MYRGP 350100W21RA	3.50	3.430	3.500	3.570	5.50	148	212	276
MYRGP 210100W21RA	2.10	2.058	2.100	2.142	3.50	140	200	260	MYRGP 355100W21RA	3.55	3.479	3.550	3.621	5.50	151	215	280
MYRGP 215100W21RA	2.15	2.107	2.150	2.193	3.58	140	200	260	MYRGP 360100W21RA	3.60	3.528	3.600	3.672	5.50	153	218	284
MYRGP 220100W21RA	2.20	2.156	2.200	2.244	3.67	140	200	260									

■ ELECTRICAL CHARACTERISTICS

● SPEC Table (V_{OUT}, t_{ONmin})
 MYRGP-B; PWM/PFM Control

Part Number	NOMINAL OUTPUT VOLTAGE	V _{OUT}			t _{ONmin}				Part Number	NOMINA L OUTPUT VOLTAGE	V _{OUT}			t _{ONmin}			
					f _{OSC} =3.0MHz									f _{OSC} =3.0MHz			
		<E-1>	<E-2>	<E-3>	<C-1>	<E-5>	<E-6>	<E-7>			<C-1>	<E-5>	<E-6>	<E-7>			
	V _{OUT(T)}	MIN.	TYP.	MAX.	V _{IN}	MIN.	TYP.	MAX.		V _{IN}	MIN.	TYP.	MAX.				
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■ TYPICAL CIRCUIT



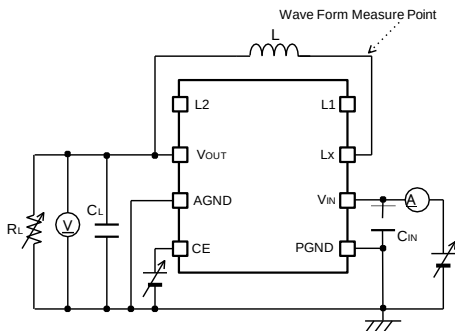
PARAMETER	VALUE
C_{IN}	10V/4.7 μ F
C_L	10V/10 μ F

NOTE:

The integrated Inductor can be used only for this DC/DC converter. Please do not use this inductor for other reasons.

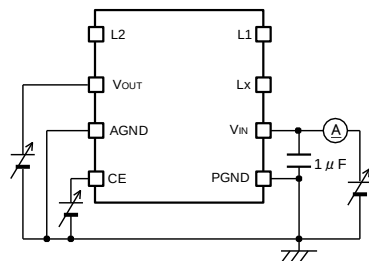
■ TEST CIRCUITS

< Circuit No.① >

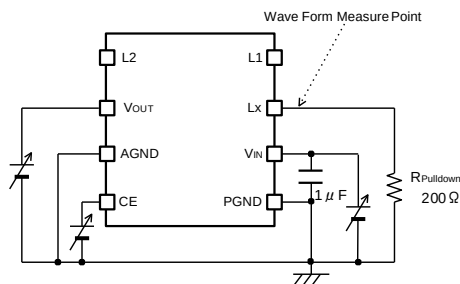


※External Components L :
1.0 μ H (Selected goods)
CIN : 4.7 μ F (Ceramic)
CL : 10 μ F (Ceramic)

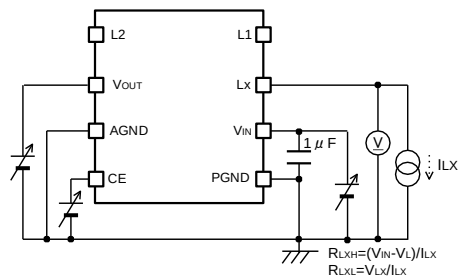
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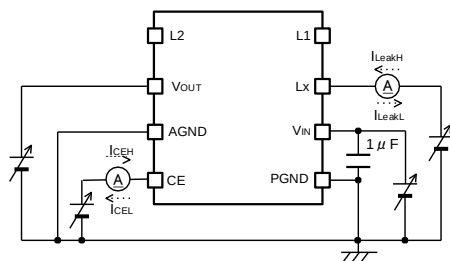
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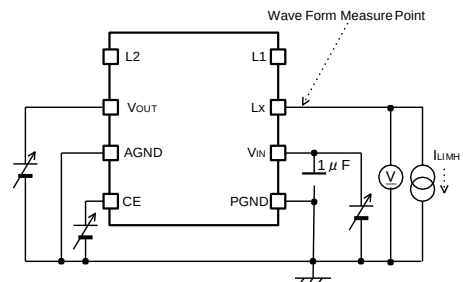
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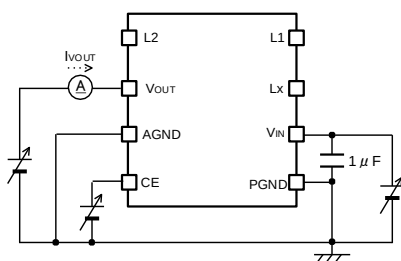
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< Circuit No.⑥ >

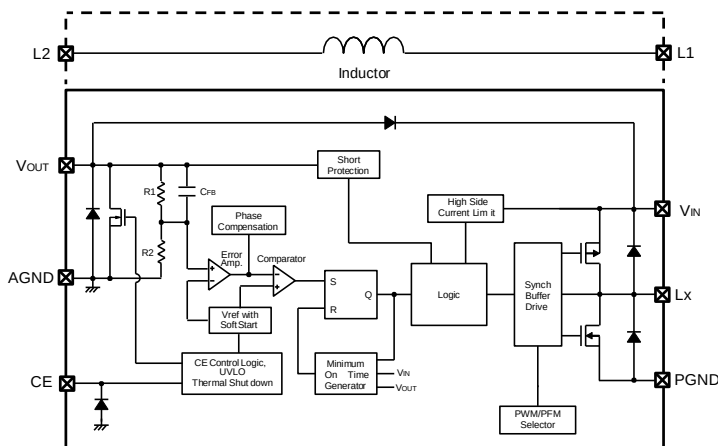


< Circuit No.⑦ >



■ OPERATIONAL EXPLANATION

The MYRGP-W/MYRGP-B series consists of a reference voltage source, error amplifier, comparator, phase compensation, minimum on time generation circuit, output voltage adjustment resistors, P-channel MOS driver transistor, N-channel MOS switching transistor for the synchronous switch, current limiter circuit, UVLO circuit, thermal shutdown circuit, short protection circuit, PWM/PFM selection circuit and others. (See the BLOCK DIAGRAM below.)



<BLOCK DIAGRAM>

The method is High Speed circuit Architecture for Transient with Constant On Time control, which features on time control method and a fast transient response that also achieves low output voltage ripple.

The on time (t_{on}) is determined by the input voltage and output voltage, and turns on the Pch MOS driver Tr. for a fixed time. During the off time (t_{off}), the voltage that is fed back through R1 and R2 is compared to the reference voltage by the error amp, and the error amp output is phase compensated and sent to the comparator. The comparator compares this signal to the reference voltage, and if the signal is lower than the reference voltage, sets the SR latch. On time then resumes. By doing this, PWM operation takes place with the off time controlled to the optimum duty ratio and the output voltage is stabilized. The phase compensation circuit optimizes the frequency characteristics of the error amp, and generates a ramp wave similar to the ripple voltage that occurs in the output to modulate the output signal of the error amp. This enables a stable feedback system to be obtained even when a low ESR capacitor such as a ceramic capacitor is used, and a fast transient response and stabilization of the output voltage are achieved.

<Minimum on time generation circuit>

Generates an on time that depends on the input voltage and output voltage (t_{on}). The on time is set as given by the equations below.

$$t_{on} (ns) = V_{OUT}/V_{IN} \times 333$$

<Switching frequency>

The switching frequency can be obtained from the on time (t_{on}), which is determined by the input voltage and output voltage, as given by the equation below.

$$f_{OSC} (MHz) = V_{OUT}(V) / (V_{IN}(V) \times t_{on}(ns))$$

<100% duty cycle mode>

When the load current is heavy and the voltage difference between input voltage and output voltage is small, 100% duty cycle mode is activated and it keeps the Pch MOS driver Tr. turning on. 100% duty cycle mode attains a high output voltage stability and a high-speed response under all load conditions, from light to heavy, even in conditions where the dropout voltage is low.

<Error amp>

The error amp monitors the output voltage. The voltage divided by the internal R1 and R2 resistors is a feedback voltage for Error Amp. and compared to the reference voltage. The output voltage of the error amp becomes higher when the feedback voltage is higher than the reference voltage. The frequency characteristics of the error amp are optimized internally.

■ OPERATIONAL EXPLANATION (Continued)

<Current Limit>

The current limiter circuit of the MYRGP-W/MYRGP-B series monitors the current flowing through the P-channel MOS driver transistor connected to the Lx pin. When the driver current is greater than a specific level, the current limit function operates to turn off the pulses from the Lx pin at any given timing. When the over current state is eliminated, the IC resumes its normal operation.

<Reference voltage source, soft-start function>

The reference voltage forms a reference that is used to stabilize the output voltage of the IC. After chip enable of the IC, the reference voltage connected to the error amp increases linearly during the soft-start interval.

This allows the voltage divided by the internal R1 and R2 resistors and the reference voltage to be controlled in a balanced manner, and the output voltage rises in proportion to the rise in the reference voltage. This operation prevents rush input current and enables the output voltage to rise smoothly.

If the output voltage does not reach the set output voltage within the soft start time, such as when the load is heavy or a large capacity output capacitor is connected, the balancing of the voltage divided by the internal resistors R1 and R2 and the reference voltage is lost, however, the current restriction function activates to prevent an excessive increase of input current, enabling a smooth rise of the output voltage.

<PWM/PFM selection circuit>

PWM control is a continuous conduction mode, and operates at a stable switching frequency by means of an on time (t_{on}) that is determined by the input voltage and output voltage regardless of the load.

PWM/PFM auto switching control is a discontinuous conduction mode at light loads, and lowers the switching frequency to reduce switching loss and improve efficiency.

The MYRGP-W series is internally fixed to PWM control.

The MYRGP-B series is internally fixed to PWM/PFM auto switching control.

<CE function>

Operation starts when “H” voltage is input into the CE pin. The IC can be put in the shutdown state by inputting “L” voltage into the CE pin. In the shutdown state, the supply current of the IC is 0 μ A (TYP.), and the Pch MOS driver Tr. and Nch MOS switch Tr. for synchronous rectification turn off. The CE pin is a CMOS input and the sink current is 0 μ A.

<UVLO>

When the V_{IN} voltage becomes 2.00V (TYP.) or lower, the P-ch MOS driver transistor output driver transistor is forced OFF to prevent false pulse output caused by unstable operation of the internal circuitry. When the V_{IN} pin voltage becomes 2.10V (TYP.) or higher, switching operation takes place. By releasing the UVLO function, the IC performs the soft start function to initiate output startup operation. The UVLO circuit does not cause a complete shutdown of the IC, but causes pulse output to be suspended; therefore, the internal circuitry remains in operation.

<Thermal Shutdown>

For protection against heat damage of the ICs, thermal shutdown function monitors chip temperature. The thermal shutdown circuit starts operating and the P-ch MOS driver and N-ch MOS driver transistor will be turned off when the chip's temperature reaches 150°C. When the temperature drops to 120°C (TYP.) or less after shutting of the current flow, the IC performs the soft-start function to initiate output startup operation.

<Short-circuit protection function>

The short-circuit protection circuit protects the device that is connected to this product and to the input/output in situations such as when the output is accidentally shorted to GND. The short-circuit protection circuit monitors the output voltage, and when the output voltage falls below the short-circuit protection threshold voltage, it turns off the Pch MOS driver Tr and latches it. Once in the latched state, operation is resumed by turning off the IC from the CE pin and then restarting, or by re-input into the V_{IN} pin.

■ OPERATIONAL EXPLANATION (Continued)

<CL High Speed Discharge>

The product can quickly discharge the electric charge at the output capacitor (CL) when a low signal to the CE pin which enables a whole IC circuit put into OFF state, is inputted via the N-ch MOS switch transistor located between the VOUT pin and the GND pin. When the IC is disabled, electric charge at the output capacitor (CL) is quickly discharged so that it may avoid application malfunction.

$$V = V_{OUT(T)} \times e^{-t/\tau}$$

$$t = \tau \ln(V_{OUT(T)}/V)$$

V : Output voltage after discharge

VOUT(T) : Output voltage

t : Discharge time

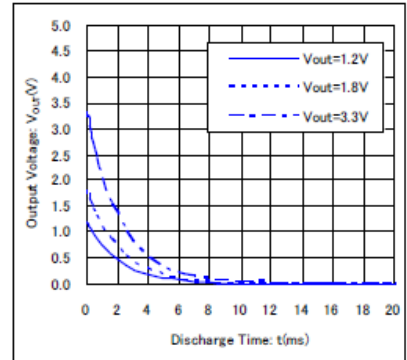
τ : CL×RDCHG

CL : Capacitance of Output capacitor

RDCHG : CL auto-discharge resistance,
but it depends on supply voltage.

Output Voltage Discharge characteristics

RDCHG=210Ω (TYP.) CL=10μF

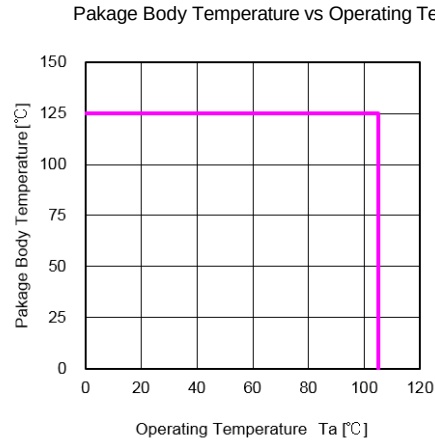
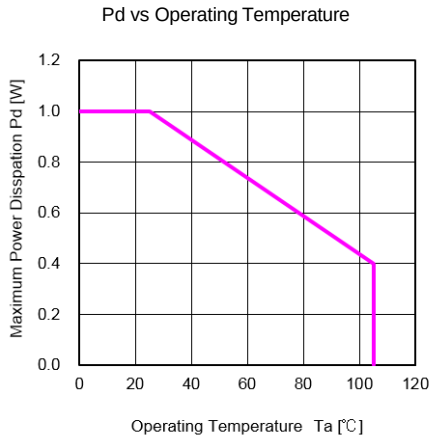


■ NOTE ON USE

1. For the phenomenon of temporal and transitional voltage decrease or voltage increase, the IC may be damaged or deteriorated if IC is used beyond the absolute MAX. specifications.
2. Spike noise and ripple voltage arise in a switching regulator as with a DC/DC converter. These are greatly influenced by external component selection, such as the coil inductance, capacitance values, and board layout of external components. Once the design has been completed, verification with actual components should be done.
3. The DC/DC converter characteristics depend greatly on the externally connected components as well as on the characteristics of this IC, so refer to the specifications and standard circuit examples of each component when carefully considering which components to select. Be especially careful of the capacitor characteristics and use B characteristics (JIS standard) or X7R, X5R (EIA standard) ceramic capacitors.
4. Sufficiently reinforce the ground wiring. In particular, reinforce near the PGND and AGND pin as fluctuations of the ground phase due to the ground current during switching may cause the operation of the IC to become unstable.
5. Mount external components as close as possible to the IC. Keep the wiring short and thick to lower the wiring impedance.
6. A feature of High Speed Transient Response -COT control is that it controls the off time in order to control the duty, which varies due to the effects of power loss between the input (V_{IN} pin) and output (V_{OUT} pin) due to the load, and thus the switching frequency fluctuates. In addition, changes in the on time due to 100% duty cycle mode are allowed. For this reason, caution must be exercised as the characteristics of the switching frequency will vary depending on the external component characteristics, board layout, input voltage, output voltage, load current and other parameters.
7. Due to propagation delay inside the product, the on time generated by the minimum on time generation circuit may not be the same as the on time that is the ratio of the input voltage to the output voltage.
8. With regard to the current limiting value, the actual coil current may at times exceed the electrical characteristics due to propagation delay inside the product.
9. The CE pin is a CMOS input pin. Do not use with the pin open. If connecting to the input or ground, use a resistor up to 1MΩ. To prevent malfunctioning of the device connected to this product or the input/output due to short circuiting between pins, it is recommended that a resistor be connected.
10. If the output voltage drops below the short circuit protection threshold voltage at the end of the soft start interval, operation will stop.
11. PWM/PFM auto switching control is a discontinuous conduction mode when the load is light, and in cases where the voltage difference between input and output is low or the coil inductance is higher than the value indicated in the standard circuit example, the coil current may reverse when the load is light, and thus pulse skipping will not be possible and light load efficiency will worsen.
12. When the input voltage is close to the minimum input voltage, the current limit circuit might not be able to work.
13. When the voltage difference between input voltage and output voltage is low, the load stability feature may deteriorate.
14. If the capacitance value is not sufficient by degrading CL due to the low temp. condition and DC bias feature, 100% duty cycle might come up for the load transient condition. Add capacitance value for CL if necessary.
15. If the capacitance value is not sufficient by degrading CL due to the low temp. condition and DC bias feature, the duty cycle might not be stable. Add capacitance value for CL if necessary.
16. MURATA places an importance on improving our products and their reliability. We request that users incorporate fail-safe designs and post-aging protection treatment when using MURATA products in their systems.

■ NOTE ON USE (Continued)

17. Please use within the power dissipation range below. Please also note that the power dissipation may changed by test conditions, the power dissipation figure shown is PCB mounted.



the power loss of mini DC/DC according to the following formula:

$$\text{power loss} = V_{\text{OUT}} \times I_{\text{OUT}} \times ((100/\text{EFFI}) - 1) \quad (\text{W})$$

V_{OUT} : Output Voltage (V)

I_{OUT} : Output Current (A)

EFFI : Conversion Efficiency (%)

Measurement Condition (Reference data)

Condition: Mount on a board

Ambient: Natural convection

Soldering: Lead (Pb) free

Board: Dimensions 40 x 40 mm (1600 mm² in one side)

Copper (Cu) traces occupy 50% of the board area

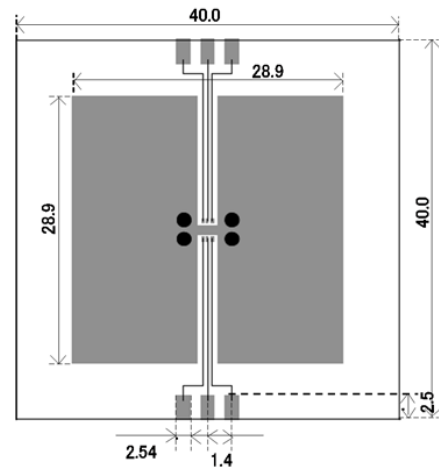
In top and back faces

Package heat-sink is tied to the copper traces

Material: Glass Epoxy (FR-4)

Thickness: 1.6mm

Through-hole: 4 x 0.8 Diameter



Evaluation Board (unit: mm)

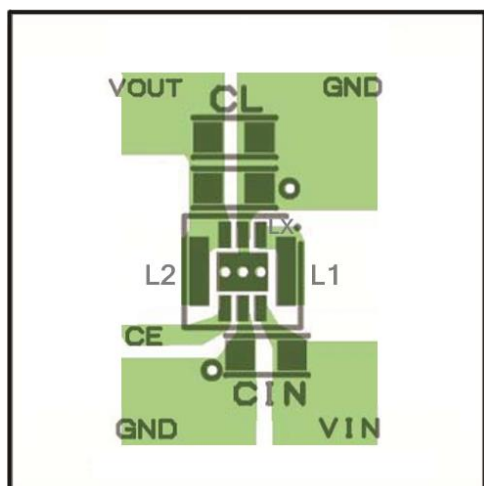
■ NOTE ON USE (Continued)

18. Instructions of pattern layouts

The operation may become unstable due to noise and/or phase lag from the output current when the wire impedance is high, please place the input capacitor(C_{IN}) and the output capacitor (C_L) as close to the IC as possible.

- (1) In order to stabilize VIN voltage level, we recommend that a by-pass capacitor (C_{IN}) be connected as close as possible to the VIN pin, PGND pin.
- (2) Please mount each external component as close to the IC as possible.
- (3) Wire external components as close to the IC as possible and use thick, short connecting traces to reduce the circuit impedance.
- (4) Make sure that the GND traces are as thick as possible, as variations in ground potential caused by high ground currents at the time of switching may result in instability of the IC.
- (5) This series' internal driver transistors bring on heat because of the output current and ON resistance of P-channel and N- channel MOS driver transistors. Please consider the countermeasures against heat if necessary.

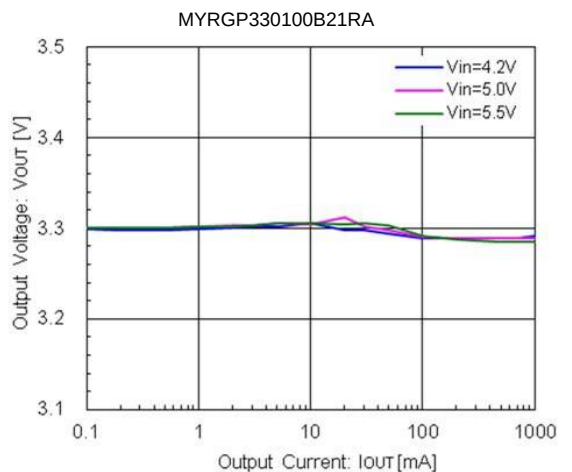
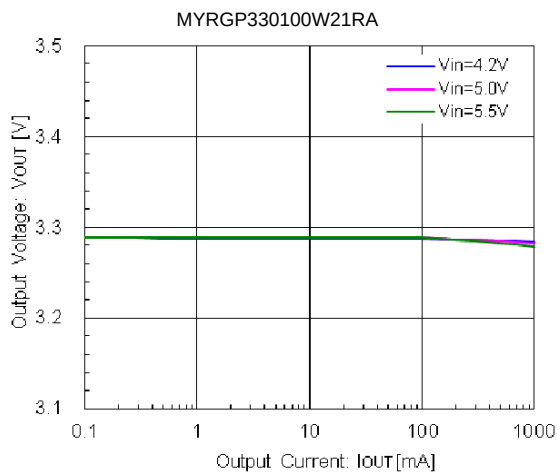
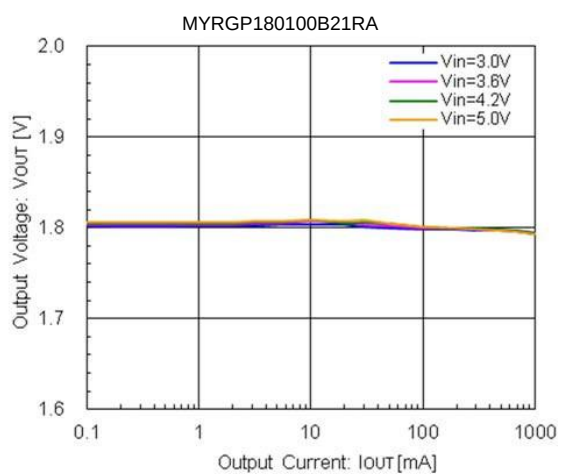
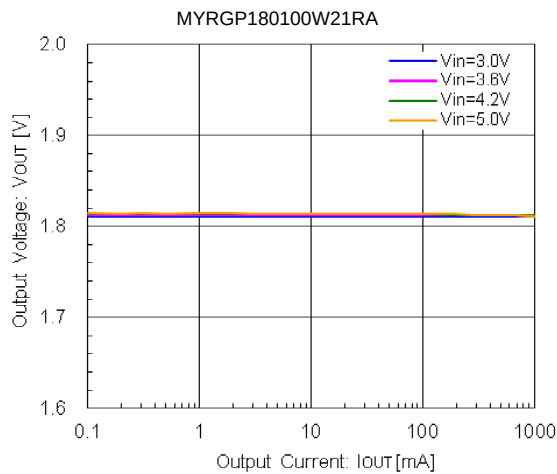
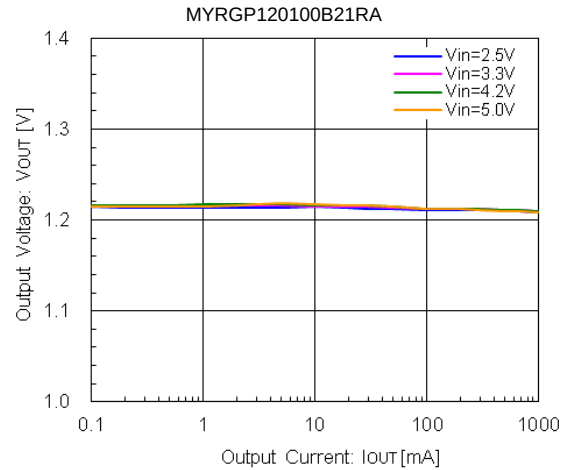
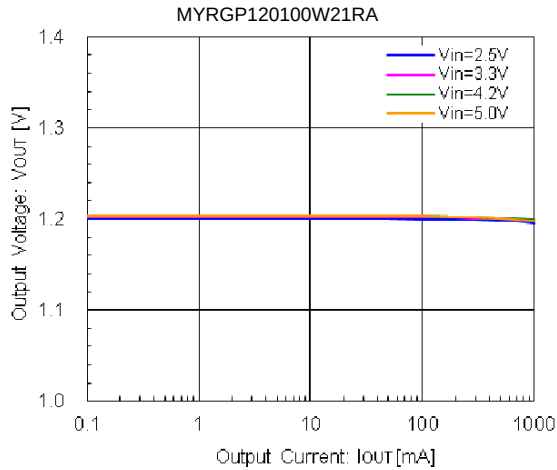
<Reference pattern layout>



<TOP VIEW>

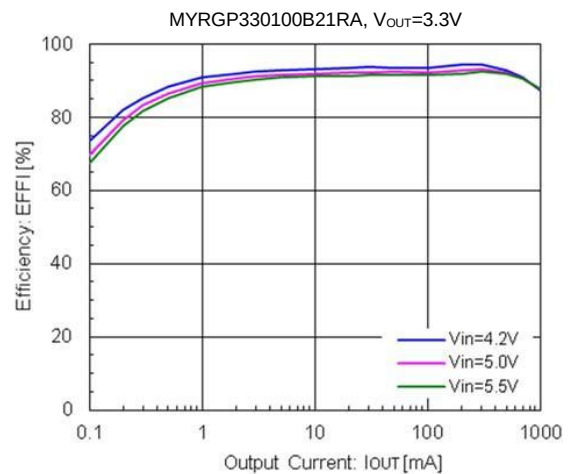
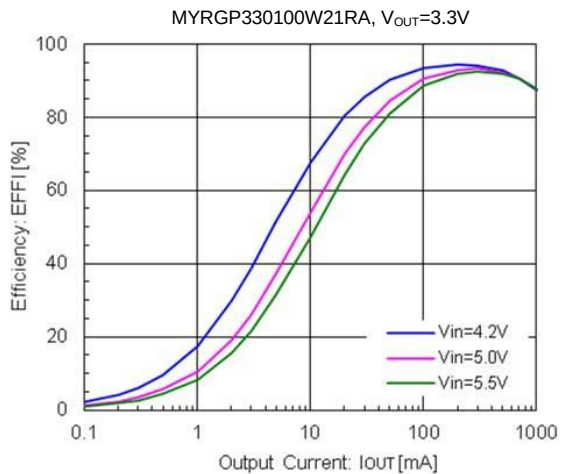
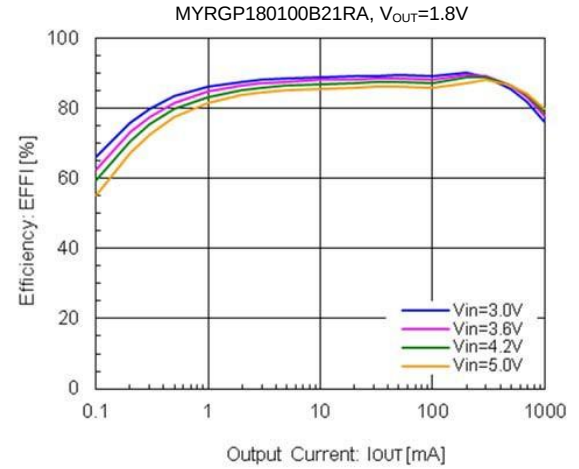
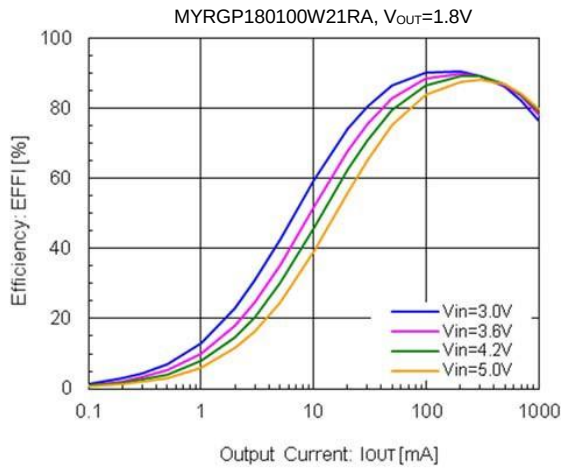
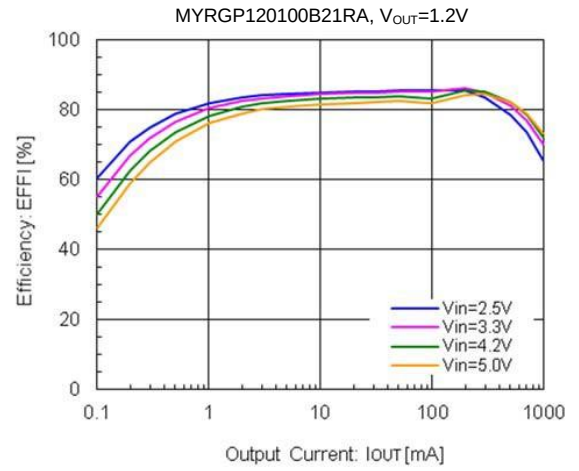
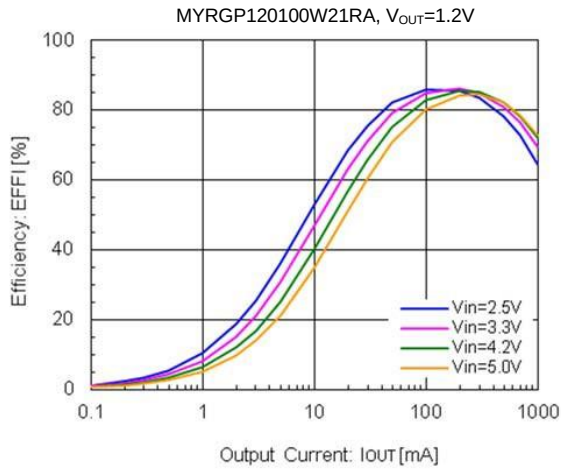
■ TYPICAL PERFORMANCE CHARACTERISTICS

(1) Output Voltage vs. Output Current



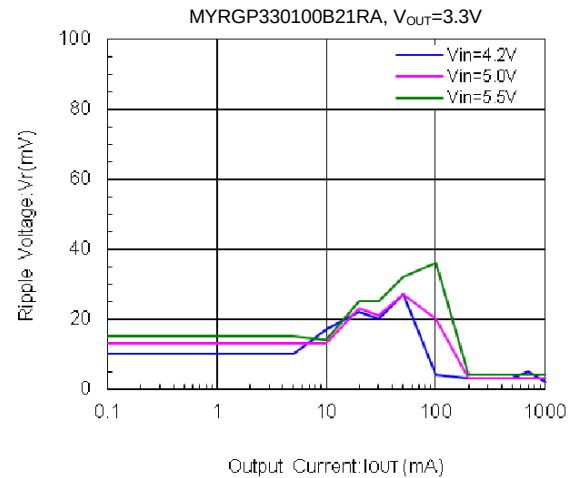
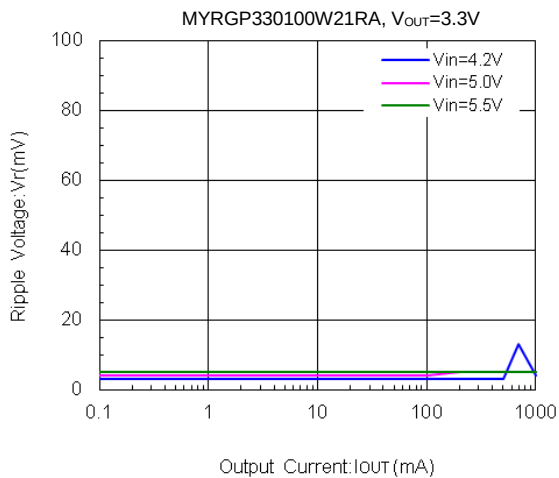
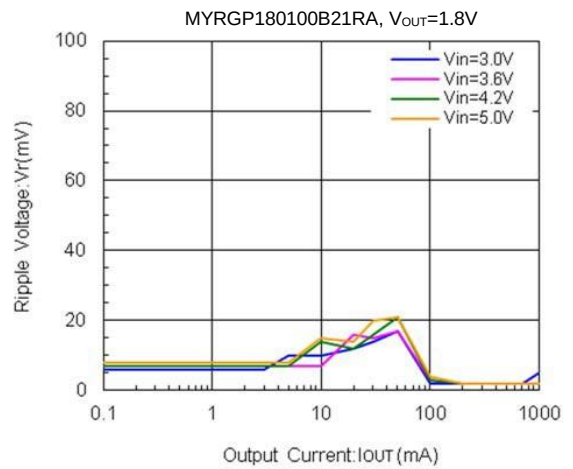
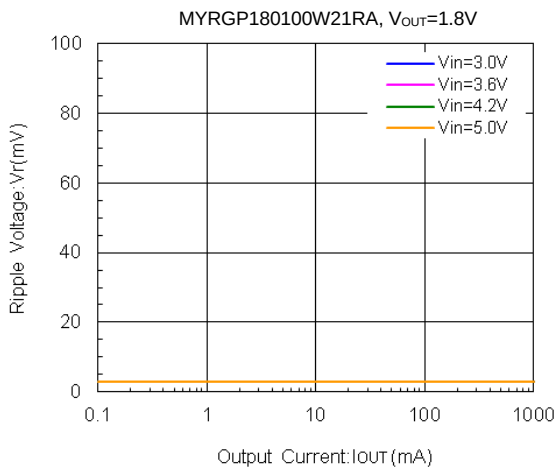
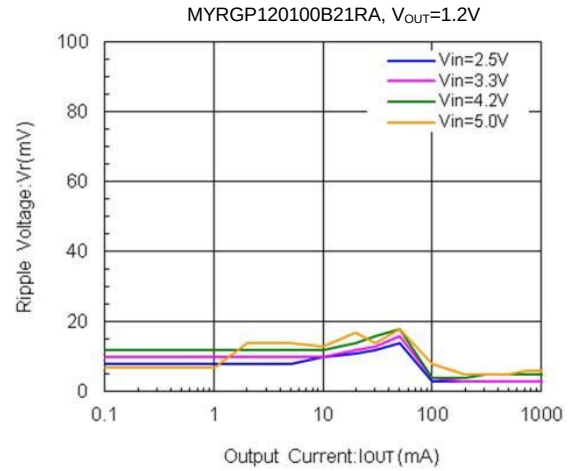
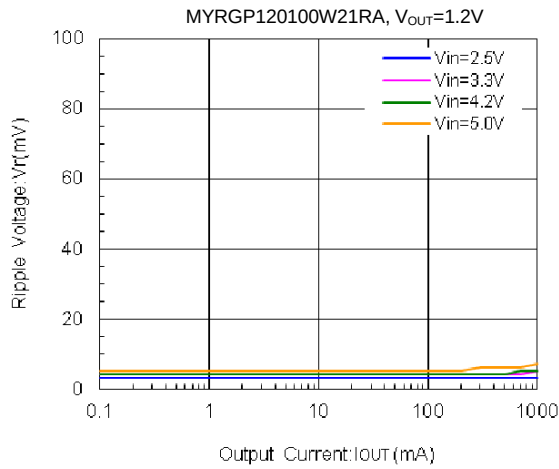
■ TYPICAL PERFORMANCE CHARACTERISTICS (Continued)

(2) Efficiency vs. Output Current



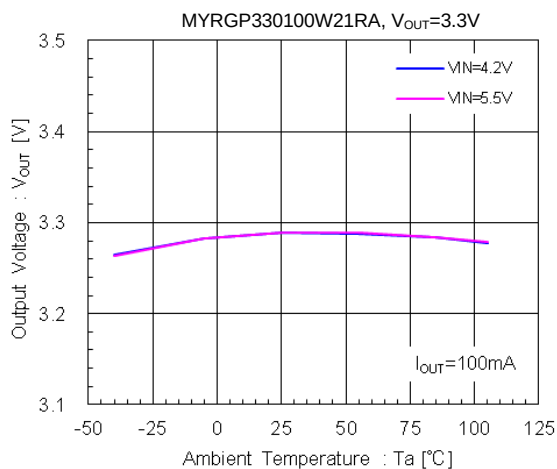
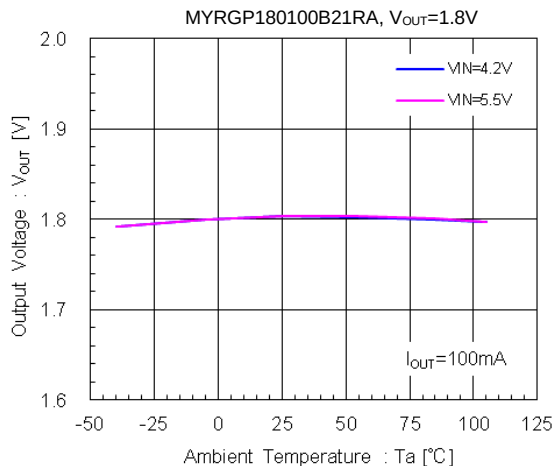
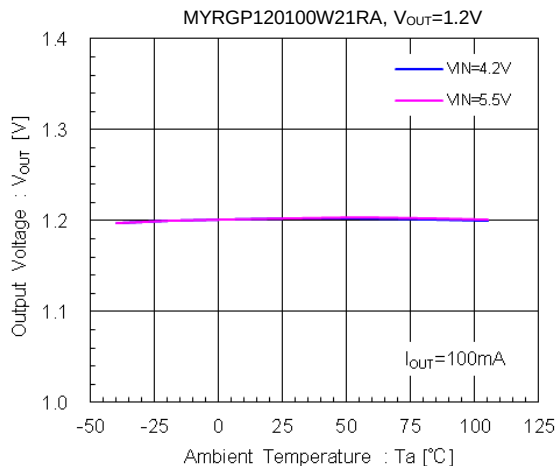
■ TYPICAL PERFORMANCE CHARACTERISTICS (Continued)

(3) Ripple Voltage vs. Output Current

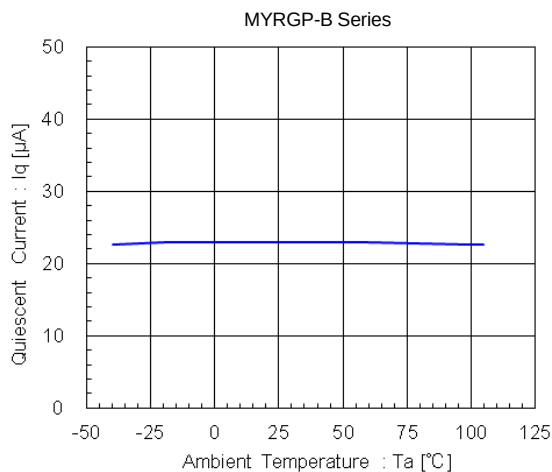


■ **TYPICAL PERFORMANCE CHARACTERISTICS (Continued)**

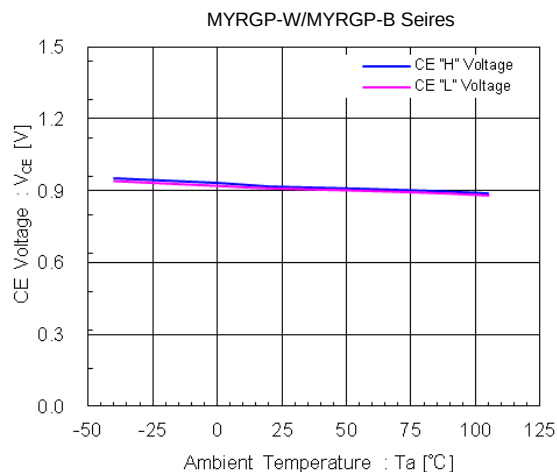
(4) Output Voltage vs. Ambient Temperature



(5) Quiescent Current vs. Ambient Temperature



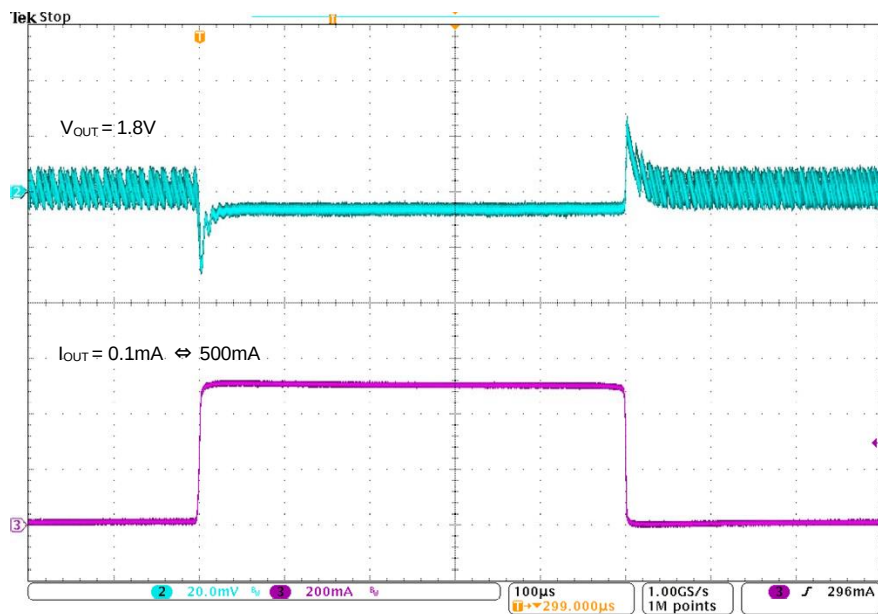
(6) CE Voltage vs. Ambient Temperature



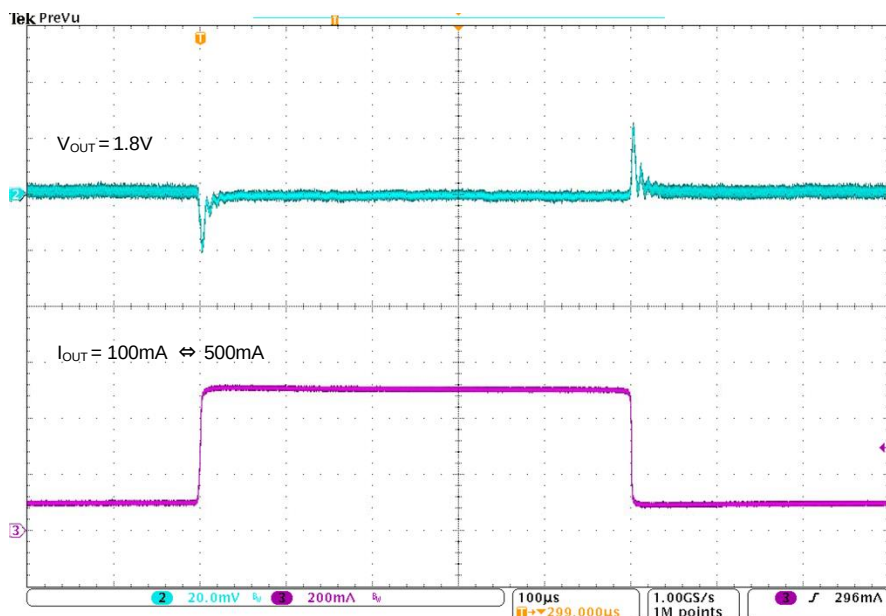
■ TYPICAL PERFORMANCE CHARACTERISTICS (Continued)

(7) Load Transient Response

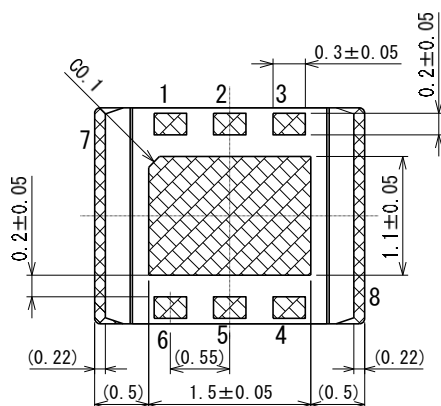
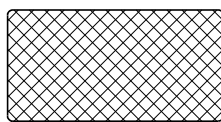
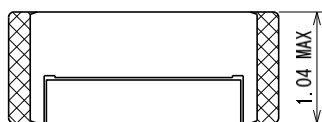
(1) MYRGP180100B21RA, $V_{IN}=4.2V$, $V_{OUT}=1.8V$ / $I_{OUT}=0.1mA \leftrightarrow 500mA$



(2) MYRGP180100B21RA, $V_{IN}=4.2V$, $V_{OUT}=1.8V$ / $I_{OUT}=100mA \leftrightarrow 500mA$



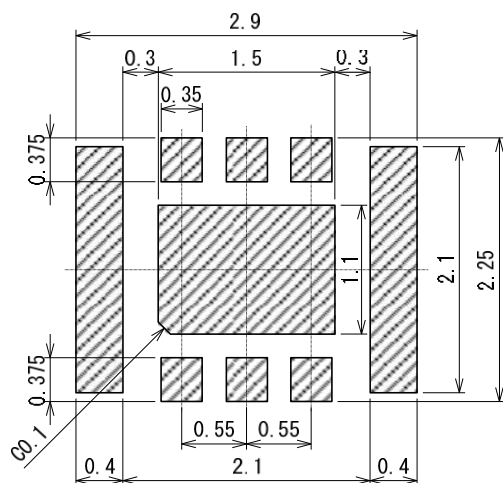
●Packaging (2.0mm × 2.5mm, h=1.0mm)



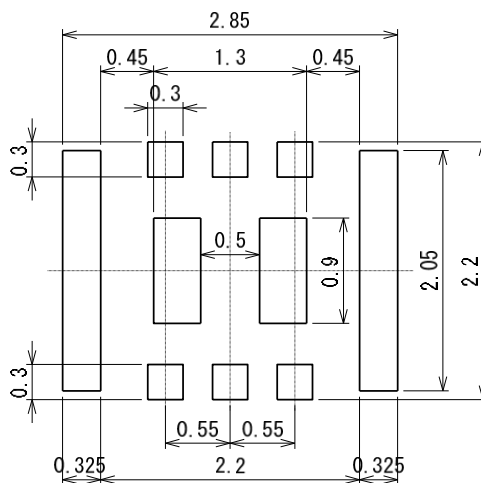
■ External Lead



●Reference Pattern Layout (unit:mm)



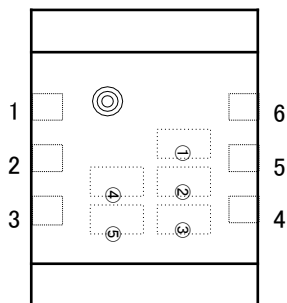
●Reference Metal Mask Design (unit:mm)



* Implementation is recommended within accuracy 0.05mm / 0.05mm.

“PicoBK™” MYRGP-W／MYRGP-B series

1.0A Inductor Built-in Step-Down DC/DC Converters



① represents products series

MARK	PRODUCT SERIES
2	MYRGP*****W21RA
3	MYRGP*****B21RA

② represents integer and oscillation frequency of the output voltage

MARK	OUTPUT VOLTAGE(V)	SWITCHING FREQUENCY (MHz)	PRODUCT SERIES
N	0.x	3	MYRGP0**100*21RA
P	1.x		MYRGP1**100*21RA
R	2.x		MYRGP2**100*21RA
S	3.x		MYRGP3**100*21RA

③ represents the decimal part of output voltage

OUTPUT VOLTAGE(V)	MARK	PRODUCT SERIES
X.0	0	MYRGP*00100*21RA
X.05	A	MYRGP*05100*21RA
X.1	1	MYRGP*10100*21RA
X.15	B	MYRGP*15100*21RA
X.2	2	MYRGP*20100*21RA
X.25	C	MYRGP*25100*21RA
X.3	3	MYRGP*30100*21RA
X.35	D	MYRGP*35100*21RA
X.4	4	MYRGP*40100*21RA
X.45	E	MYRGP*45100*21RA
X.5	5	MYRGP*50100*21RA
X.55	F	MYRGP*55100*21RA
X.6	6	MYRGP*60100*21RA
X.65	H	MYRGP*65100*21RA
X.7	7	MYRGP*70100*21RA
X.75	K	MYRGP*75100*21RA
X.8	8	MYRGP*80100*21RA
X.85	L	MYRGP*85100*21RA
X.9	9	MYRGP*90100*21RA
X.95	M	MYRGP*95100*21RA

④,⑤ represents production lot number

01~09、0A~0Z、11~9Z、A1~A9、AA~AZ、B1~ZZ in order.

(G, I, J, O, Q, W excluded)

Note: No character inversion used.